

Synchronizer for RZ-WDM Signals and method for synchronization

FIELD OF INVENTION

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The present invention relates to a synchronizer for RZ (Return to Zero) modulated WDM (wavelength division multiplex) signals.

- 10 The invention also relates to a method to synchronize the RZ-WDM signals with the help of a variable delay line and an electrical feedback loop.

15 Background of the Invention

- Demand for broadband services (such as high quality data transfer, high definition television and video conferencing) may require telecommunications networks to operate with TBit/s capacities by
- 20 the first decade of the next century. In order to meet this capacity demand, all-optical or "transparent" networks have been proposed, which networks employ either high speed optical time division multiplexing (OTDM) or wavelength division multiplexing (WDM) to attain the high data-rate. The transparent optical networks rely on
- 25 optical switching and routing to maintain a transparent path between the source and destination nodes.

- In transmission systems, electronic clock recovery circuits are generally used, conventional techniques using electronic filtering in
- 30 the post detection circuitry. For instance, a high Q electrical filter may be used to extract the clock component in a received data modulation spectrum.

Within transparent optical network architectures, electronic clock recovery techniques are well known. Using a WDM transmission

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scheme a large number of independent wavelength channels are transmitted. Each channel is modulated independently from the adjacent channels. The channels are not or only at the transmitters synchronized to each others. The regeneration function during a transmission line includes a re-modulator. If the channels are not synchronized at the regeneration stage the number of regenerators is equal to the number of channels at each regeneration stage.

In return-to-zero (RZ) coding, the frequency spectrum of a coded signal will include a strong peak at the clock frequency. Clock extraction can then be achieved by filtering at the clock frequency and rectifying the result. This involves signal conversion to electronic form. Afterward the clock signals are synchronized by using a phase comparing function. This solution is limited to lower bit rates or increase costs for the requested electronic circuit.

A regeneration and synchronization is known from the US 6,028,898. The signal regenerator comprises a threshold adjustment circuit; a phase adjustment circuit and a re-timing circuit.

Otherwise a complete clock recovery of all RZ-WDM signals is not needed in every regeneration stage of a transmission line. So the invention is to synchronize the WDM channels in an easy way, without a clock recovery in each channel.

The invention is realized in

A synchronizer for more than one optical RZ signal in a wavelength multiplex transmission system comprising

- at least one variable delay line (1) with an input receiving RZ-WDM optical channels (2)
- at least one delay controller (3) receiving the RZ-WDM output optical channels (10)

- the delay controller generating a control signal depending on the power of the RZ-WDM output optical channels (10) and
 - a control circuit (9) to control the at least one delay line (1) in such a way that the optical WDM channels are synchronized.
- 5 The invention comprises also a method for synchronization of RZ-WDM optical signals realized by the steps:
- Separating two channels from the WDM multiplex
 - Synchronizing them by
 - Analyzing the power of the two channels
- 10
- Generating a control signal for the variable delay line
 - Controlling the delay line
 - And feeding the resulting synchronized signals back to the next subset of channels so that the synchronized channels are one of the two channels of the subset.

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BRIEF DESCRIPTION OF THE DRAWINGS

A synchronizer will now be described as an example of an embodiment of the present invention, with reference to the

20 accompanying drawings, in which:

FIG. 1 shows channel power over time

FIG. 2 shows a principle structure of the synchronizer

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FIG. 3 shows a first embodiment of the synchronizer

FIG. 4 shows a second embodiment of the synchronizer.

Fig. 1 shows the function of the resulting power of two WDM

30 channels over a certain delay DT.

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To understand the dependence of this function we consider two WDM channels modulated in RZ Signals. This optical RZ-WDM channels are named a and b. The two channels are shifted in time by a time period of DT measured in Tbit. The modulation rate for this two channels is (1/Tbit) bit/s. With this modulation scheme the two channels a and b both have a component at (1/Tbit) Hz in their electrical spectrum of power Pa and Pb. If the channels a and b are simultaneously detected in a photo detector it can be shown that the power of the component at the frequency (1/Tbit) Hz in the electrical spectra of the optical sum of two incoherent channels follows the function:

$$P = \sqrt{(P_a - P_b)^2 + 4P_a P_b \cdot \cos^2\left(\frac{\pi * DT}{T_{Bit}}\right)}$$

If Pa= Pb the relation is:

$$P = 2P_a \left| \cos\left(\frac{\pi * DT}{T_{Bit}}\right) \right|$$

It can be seen that the power P of the electrical component at frequency (1/Tbit) Hz of the optical sum of the two channels is maximal when the pulsed of both channels are synchronized and that P is minimal when the pulsed of both channels are out of synchronization.

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Fig. 2 shows in a schematic way the synchronizer for two optical channels a and b. The optical RZ-WDM channels 2 are connected with a first port of a variable delay line 1. The output of the a variable delay line 1 is connected to a transmission line for the synchronized RZ-WDM channels 10. A tap 14 splits optical power and is connected to a delay detector 3. The delay detector 3 is attached to a control system 9 and this control system 9 is attached

to a second input port of the variable delay line 1. The delay detector 3 consists of a photo detector 5 connected to the input and the tap 14. The photo detector is linked to a band pass filter 6 which is linked to a HF-powermeter 7. The powermeter 7 is
 5 attached to a decision circuit 8 and the output connected to the control system 9.

The input RZ WDM channels 2 are not synchronized The shift between the two wavelength channels is DT . The distance between
 10 the digital signals for example the signals "1" in the figure is T_{bit} . This signals are fed to the variable delay line. What is need is a control signal to initiate the active shift of one channel compared to the other. At the tap 14 a part of the optical signal, that is not influenced by the delay line in the start of the procedure, is taken
 15 to feed the delay detector. Herein the photo detector receives the optical signal changing it in an electrical broadband signal. This electrical signal is then filtered by a band pass filter centered around the data clock frequency to delete side band and noise signals. The filtered signal is analyzed by a HF power meter.

20 This power meter (it could be a RF detection diodes) sensitive to the RF clock level converts the HF signal to a DC electrical level compatible with the so-called decision element 8. The decision element could be realized using electronic enslavement based on
 25 operational amplifiers circuitry and transforms the received information in a signal compatible with the variable optical delay line control system 9. A deviation from the maximum RF level is detected in the decision circuit as an error voltage which is lowered using a P.I.D. regulation technique (Proportional, Integral,
 30 Derivative) driving the control system.

For this embodiment the variable delay line is wavelength dependent.

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In another embodiment the decision circuit look for the minima of the signals . The signal structure is also flexible.

The optical delay line could be an adjustable air gap electrically controlled with a step-motor to change the optical path.

Fig. 3 shows an embodiment of the invention for more than 2 channels. The explanation uses three channels without limiting the scope of the invention.

- 10 The input RZ-WDM channels 2 are connected to a circulator 12. The first port of the circulator 12 is connected with a first tap 14a. The tap 14a is also link over a first fiber grating 11a to a first variable delay line 1a. The output of the delay line 1a is over a second tap 14b and a second grating filter 11b connected to a second delay
- 15 line 1b.

The second delay line is linked to a third grating filter 14c. Each tap 14a and 14b is attached over a delay detector 3a, 3b and a control circuit 9a, 9b with an input port of a delay line 1a, 1b.

- 20 The fiber gratings 11 are reflecting one wavelength of the WDM scheme. The fiber grating 11c is reflecting λ_3 . So in the variable delay line 1b λ_3 and λ_2 are compared and synchronized. For the two channels are then synchronized the variable delay line 1a shift $\lambda_3 = \lambda_2$ versus λ_1 . The resulting synchronized signal is fed in the
- 25 circulator 12 and transmitted over a second port of the circulator 12.

- With this embodiment the RZWDM channels are synchronized beginning with channel N. Than channel N and (N-1) are synchronized , than channel N-2 with the synchronized channels N
- 30 and N-1 and finally the channel 1 is synchronized with the already synchronized channels (2....N).

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Another preferred embodiment is shown in figure 4. In this embodiment the circulator 12 is connected over a grating filter 11a with the variable delay line 1a. The tap 14a connects the second port of the circulator to the variable delay line 1a. The link contains
 5 an optical filter 13a for λ_1 and λ_2 the delay detector 3a and the control circuit 9a.

The next stage of the synchronizer with variable delay line 1b is also connected over a tap to the second output line of the circulator 12. This stage contains an optical filter for the wavelengths λ_2 and
 10 λ_3 . So every stage of the synchronizer synchronize two adjacent channels.

In another embodiment another filter structure for the optical filters is used. The filters filter λ_1 and one additional wavelength out of the WDM. This filter design allows a synchronization of each
 15 channel with channel 1.

The synchronizer is used in an intensity/phase modulator in a regeneration stage of the transmission line. For this purpose a high quality band pass filter is used with $Q > 1000$ to get an efficient clock recovery system and for exactly driving the modulator. The
 20 other channels can be synchronized using a cheaper low quality filter for example with $Q = 100$.

The whole system is a feed back control loop with a high tolerance versus changes in the optical input power and versus dispersion effects. For the synchronizer is adjusting the delays between
 25 channels automatically a slight temperature shift on the fiber or a modification of the fiber birefringence and as a result the group velocity of the signal in a channel is leveled out.

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